

Developing Laser Spot Position Determination Circuit Modeling and Measurements with a Quad Detector

Mohamed Fathy Heweage, Xiao Wen, and Ahmed Eldamarawy

Abstract—In this paper we are interested in modeling, simulation, and measurement of laser beam spot position determination circuit (LSPDC) with a quadrant detector (QD). The photodiode is modeled using its differential equations then implemented using ORCAD –Pspise. Starting with the output of the detector, the circuit modeling and simulation for the preamplifier, post amplifier, ADC and microcontroller card were performed. PCB card implementation was performed and integrated with the QD and the necessary optical system components. The implementation of the full LSPDC with the detector is performed. General system for testing the spot LSPDC was performed with a laser source simulating the incident radiation and a mechanical system to enable adjusting different angles. Good agreement has been achieved between the measured data and the adjusted angles with a small error attributed to the error in the adjusted mechanical system and the stability of the used laser source.

Index Terms—Photodiode modeling, quad detector, spot position measurements.

I. INTRODUCTION

Laser tracking systems (LTS) are used for detecting and tracking the targets by obtaining the position information of these targets. The main parts of the LTS are the optical system for collimating the reflected beam from the target, the position sensitive detector which converts the incident laser spot to its corresponding photocurrent, and the electronic circuits that, filter and convert the photocurrent to its corresponding voltage. The position sensitive diode (PSD) is the core component in the laser tracker. A PSD is a continuous silicon photodetector used for optical position sensing and basically consists of a uniform resistive layer, which is formed on a silicon substrate. A pair of electrodes is formed at the ends of the resistive layer from which photocurrents are measured. These photocurrents are generated as a result of the incident radiation effect and their magnitude is relative to the distance of the electrode to the center of the beam spot on the sensor's active area. The photocurrents are typically amplified and converted to form the measured voltage signal used by the control system. PSDs can be divided into two general groups: segmented PSDs and lateral effect PSDs in this paper we will consider segmented PSDs which are quadrate PSD (QD), each is divided into four

segments, separated by a small gap, called the dead region. In QD the position is extracted by comparing each segment's photocurrent [1], [2].

In next sections, we will cover the main components of LSPDC including a preamplifier, post amplifier, inverse amplifier and the analog to digital converter (ADC) in addition to the QD. And then we will discuss the implementation of the full position determination circuit with the detector. Also, we will show how the microcontroller was used to get the output from the ADC, and perform the processing on the signal to calculate the position information by the program that shows the position graphically on the computer display.

II. LASER POSITION DETERMINATION CIRCUIT

Fig. 1 shows the general block diagram of the four quadrant detector with a processing circuit for laser position determination. Starting from the output of the quadrant photodiode detector, the output of the photodiode is a current and with a very small value (with range in μA), all the processing made on the signal treats the signal as a volt not as a current (due to that the used instruments and equipment work on the voltage) [3], [4].

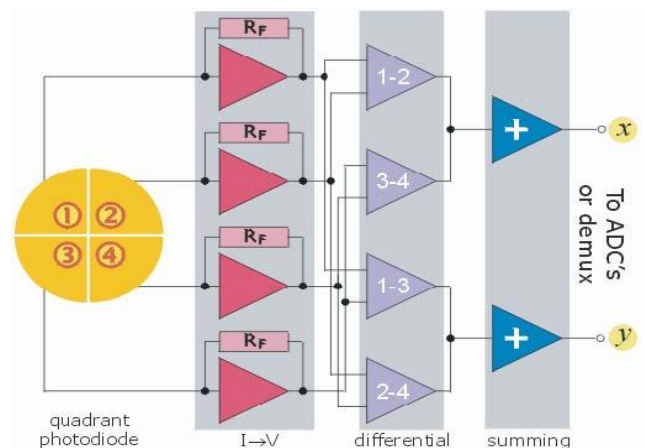


Fig. 1. General block diagram for the four quadrant detector with processing circuit.

We can overcome this problem with the usage of the trans-impedance (TI) amplifier which transfers the output current from the photodiode into an equivalent value of the volt. The low-level current can be raised by using the trans-impedance amplifier (TIA) which is used to transfer and amplify the detector output current into the volt. First stage of Fig. 1 is the quadrant photodiode, in the second stage the photodiode output current is converted to the volt with the use of the TIA, the third stage is used to differentiate

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the four signals volts from each other, and in the last stage the summing of the signals is done to calculate the spot position, this is done using the following equations [4,5,6]:

$$X = K_Q \frac{(W_1 + W_3) - (W_2 + W_4)}{W_1 + W_2 + W_3 + W_4} \quad (1)$$

$$Y = K_Q \frac{(W_1 + W_2) - (W_3 + W_4)}{W_1 + W_2 + W_3 + W_4} \quad (2)$$

where X and Y represent the position coordinate of the spot, 1,2,3,4 are the four signals from the four photodiodes as shown in Fig. 1, K is the proportional factor for the quadrant detector.

A. Preamplifier Stage

Firstly we want to convert the photodiode output current to the volt this is used by the TIA. The preamplifier is the first stage in the system and is considered to be the most critical part. We used the integrating circuit to transforms current pulse into a voltage pulse with sufficiency relative to a charge conveyed by the present current pulse. In general, the preamplifier function is to amplify a low-level signal to the wanted level. The preamplifier provides voltage gain without significant current gain. The basic form of the TIA is the noninverting operational amplifier with the feedback resistance and capacitance as shown in Fig. 2 (a), the feedback resistance for the gain amplification value and the feedback capacitor is used to prevent the high-frequency oscillation and noise, also act as a low-pass filter [7], [8], [10].

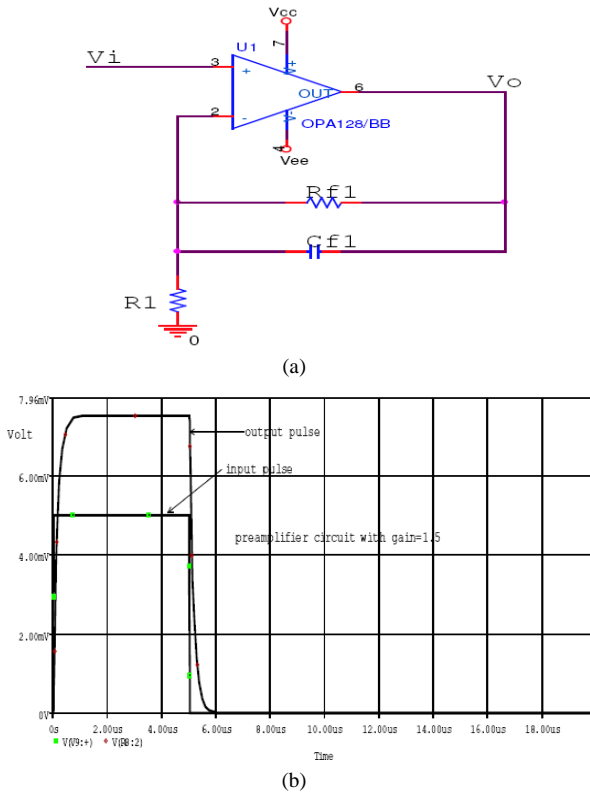


Fig. 2. (a) General schematic of the non-inverting operational (b) The output of the preamplifier with the input pulse.

From the circuit shown in Fig. 2 we can determine the gain

of the circuit by the equation:

$$\frac{V_o}{V_i} = G = \left(1 + \frac{R_{f1} // C_{f1}}{R_1} \right) \quad (3)$$

$$G = \left(1 + \frac{\frac{R_{f1} / SC_{f1}}{R_{f1} + 1 / SC_{f1}}}{R_1} \right) = \left(1 + \frac{R_{f1} / SC_{f1}}{R_1 R_{f1} + R_1 / SC_{f1}} \right) \quad (4)$$

$$G = \left(1 + \frac{R_{f1}}{R_1 + R_1 R_{f1} SC_{f1}} \right) = \left(1 + \frac{R_{f1} / R_1}{1 + R_{f1} SC_{f1}} \right) \quad (5)$$

for the DC gain $S=0$ so that the gain will be

$$G = \left(1 + \frac{R_{f1}}{R_1} \right) \quad (6)$$

Fig. 2b) shows the calculated output of the preamplifier with the gain amplifier $G=1.5$.

Gain circuit stage

The second stage circuit (gain circuit) is used to increase the level of the output volt from the preamplifier because the preamplifier has a limited amplification level.

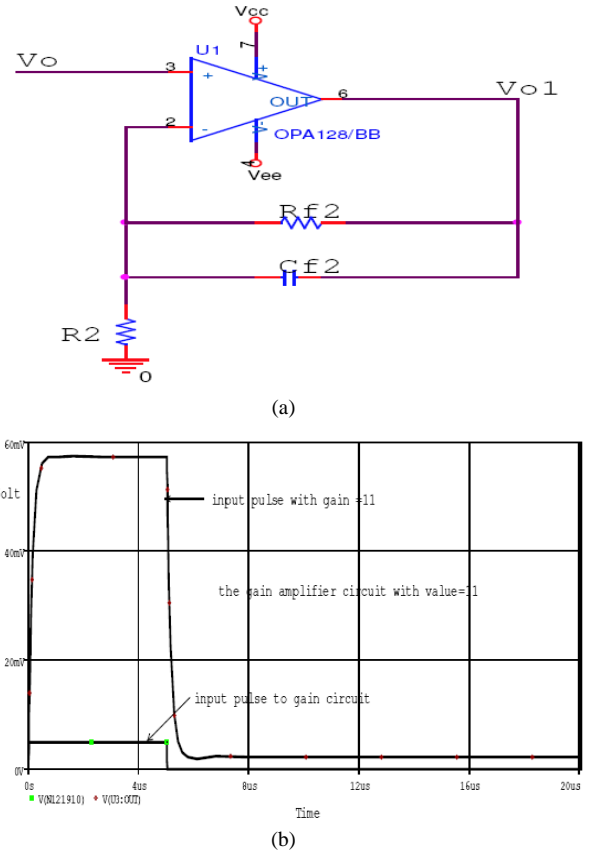


Fig. 3. A general schematic of the non-inverting operational amplifier with the gain (a), and (b) the output of the gain amplifier with the input pulse.

By using the noninverting operational amplifier shown in Fig. (3-a), for the gain value determined according to the value of feedback resistance according to the basic equation

of the operational amplifier [3], [10]. We can determine the gain of the circuit by the equation:

$$\frac{V_{o1}}{V_o} = \left(1 + \frac{R_{f2}}{R_2} \right) \quad (7)$$

By substituting from equation (5) into equation (7) we have to produce the equation which provides the summing of the two circuit as follows:

$$V_{o1} = V_i \left(1 + \frac{R_{f1}/R_1}{1 + R_{f1}SC_{f1}} \right) \left(1 + \frac{R_{f2}}{R_2} \right) \quad (8)$$

Fig. (3-b) shows the output of the gain amplifier with the gain amplifier $G=11$, the figure shows the two curves of the input with amplitude (5mv) and the output with amplitude (55.56mv) this prove the gain is $G=11$

B. Inverse Signal Circuit Stage

The third stage is used for the inverse signal circuit (emitter follower) [3] with the gain value to be a positive sign and can be read with the analog to digital converter (ADC).

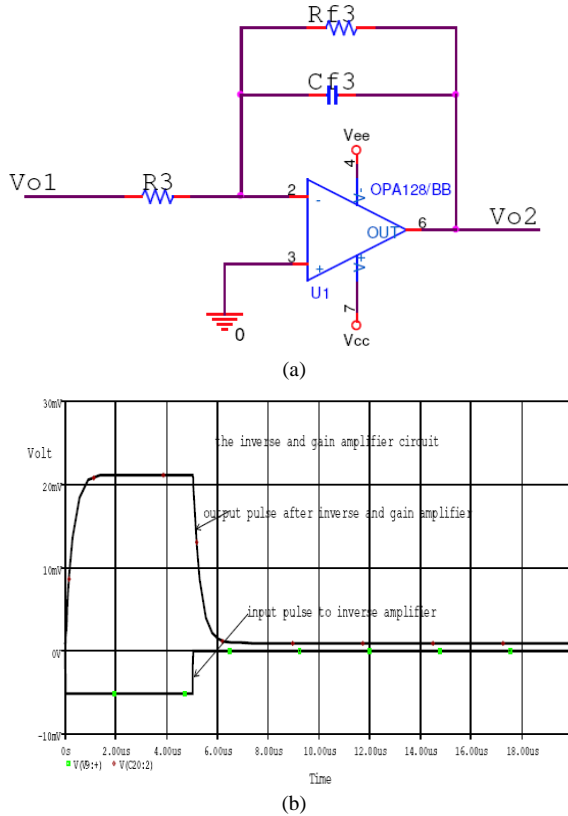


Fig. 4. A general schematic of the inverting operational amplifier with the gain amplitude (a), and (b) the output of the inverse amplifier with the input pulse.

From the circuit shown in Fig. (4-a) we can determine the gain of the circuit by the equation:

$$\frac{V_{o2}}{V_{o1}} = \left(- \frac{R_{f3} // C_{f3}}{R_3} \right) \quad (9)$$

$$\frac{V_{o2}}{V_{o1}} = \left(- \frac{\frac{R_{f3} / SC_{f3}}{R_{f3} + 1 / SC_{f3}}}{R_3} \right) = \left(- \frac{R_{f3} / SC_{f3}}{R_3 (R_{f3} + 1 / SC_{f3})} \right) \quad (10)$$

$$\frac{V_{o2}}{V_{o1}} = \left(- \frac{R_{f3}}{R_3 R_{f3} SC_{f3} + R_3} \right) \quad (11)$$

For the DC gain $S=0$ so that the gain will be

$$\frac{V_{o2}}{V_{o1}} = \left(- \frac{R_{f3}}{R_3} \right) \quad (12)$$

By substituting from equation (8) (summing circuit) into equation (11) we have to produce the equation which provides the summing of the two circuits as follows:

$$V_{o2} = -V_i \left(1 + \frac{R_{f1}/R_1}{1 + R_{f1}SC_{f1}} \right) \left(1 + \frac{R_{f2}}{R_2} \right) \left(\frac{R_{f3}}{R_3 + R_3 R_{f3} SC_{f3}} \right) \quad (13)$$

For the DC gain $S=0$ so that the gain will be the total system amplification given by:

$$V_{o2} = -V_i \left(1 + \frac{R_{f1}}{1 + R_1} \right) \left(1 + \frac{R_{f2}}{R_2} \right) \left(\frac{R_{f3}}{R_3} \right) \quad (14)$$

Fig. (4-b) shows the output of the inverse amplifier with the gain amplifier, the figure shows the two curves of the input and the output after inverse and gain amplified.

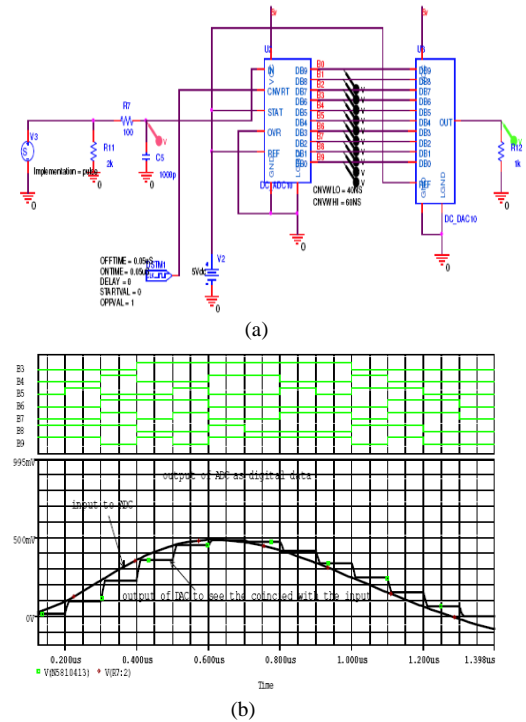


Fig. 5. Schematic diagram of the analog to digital converter (ADC) value and the digital to analog converter (DAC) (a), (b) input pulse to the ADC and the digital output of the ADC.

C. Analog to Digital Converter (ADC) Stag

Fig. (5-a) shows the circuit diagram of the analog to digital

converter (ADC) [12] and the digital to analog converter (DAC) to make sure that the input to ADC is the same as the output of the DAC. Now the output at the Vo2 will be the input to the ADC to take the value of each part of the quadrant photodiode and take the output from the analog to digital converter into the microcontroller to apply the processing on the output signal and make the determination of the laser spot position. Figure (5-b) shows the input and the output of the analog to digital converter (ADC) value and the digital to analog converter (DAC) to make sure that the input to ADC is the same as the output of the DAC, from the figure it is clear that the input to the ADC has a good agreement with the output of the DAC.

D. Laser Spot Position Determination Circuit Diagram

Now we will use the full circuit diagram for the laser spot position determination. Starting with the photodiode model [1], [2], [10], [11] and including the preamplifier circuit, the gain circuit, the inverse amplifier, and the ADC. The analog signal output from each section of the 4 quadrant detector is to be read from the microcontroller to make the processing on the signal and provide X and Y position of the laser spot.

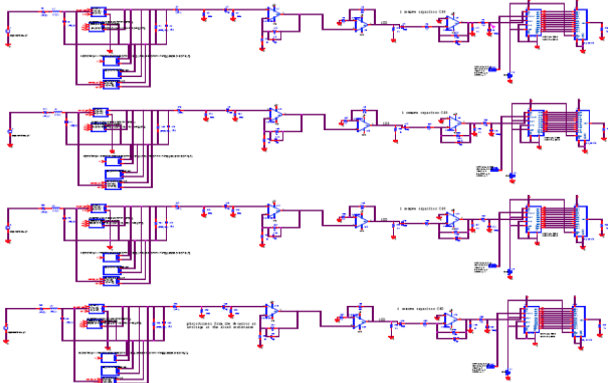


Fig. 6. The full schematic of the laser spot position determination circuit.

Fig. (6) shows the full schematic of the laser spot position determination circuit [1], [2], [11].

III. MODEL SIMULATION

To examine the system model performance, we will show the simulation of different spot position on the quadrant detector.

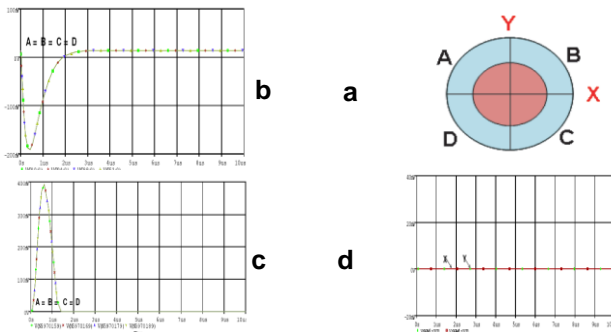


Fig. 7. laser spot at the center of QD (a), 4 PD output after amplifier (b), after inverse amplifier (c), spot position interims of X and Y (d).

Fig. 7a) shows the laser spot with the quadrant detector

when the laser spot at the center of the quadrant detector. When the spot is at the center of the quadrant detector the illuminated areas of the four photodiodes are the same, then the represented value for the incident radiation is the same for the four photodiodes. Fig. 7b) shows the output of the four photodiodes after the amplifier, the output is the same. Fig. 7c) shows the output of the four photodiodes after the inverse amplifier, the positive output is needed since the ADC works on the positive signal, for the position information we take the output from ADC and make processing on it to calculate the Y and X equations for laser spot determination according to equation (1) and (2)

Fig. 7d) shows the output value of X and Y for this condition $X=Y=0$ due to spot center with the detector center. The second case for the laser spot shifted in Y axis and zero shift in X axis the incident radiation on $C=D=2A=2B$.

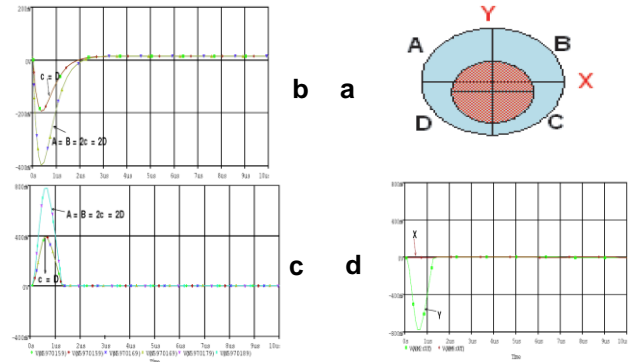


Fig. 8. laser spot shifted from the center of the quadrant detector (a), 4 PD output after amplifier (b), after inverse amplifier (c), spot position interims of X and Y (d).

Fig. 8a) show the laser spot with the quadrant detector the laser spot shifted from the center of the quadrant detector in negative Y axis and zero shift in X axis Starting with the spot position shifted in Y axis from the center of the quadrant detector in this case the illuminated areas of the four photodiodes are not the same ($A = B, C = D$), then the represented value for the incident radiation according to the illuminating area of every photodiode.

Fig. 8b) show the output of the four photodiode after the amplifier where the output of A and B is the same and the output of C and D is the same due to the incident radiation on the photodiode $A=B$ and $C=D$. Figure (8-c) show the output of the four photodiode after the inverse amplifier, the positive output due to the ADC work on the positive signal. for the position information, we take the output from ADC and make processing on it to calculate the Y and X equations for laser spot determination according to equation (1) and (2). Figure (8-d) show the output value of X and Y for this condition $X=0$ and $Y = -ve$ value due to spot center shifted in the negative Y with the detector center.

The third case for the laser spot shifted in X axis and zero shift in Y axis the incident radiation in this case $B=C=2A=2D$.

Fig. 9a) show the laser spot with the quadrant detector the laser spot shifted from the center of the quadrant detector in positive X-axis and zero shift in Y axis Starting with the spot shifted in X axis from the center of the quadrant detector, in this case, the illuminated areas of the four photodiodes are not the same and follow the condition $B=C=2A=2D$, then the

represented value for the incident radiation is not the same for the four photodiodes. Fig. 9b) shows the output of the four photodiode after the amplifier the output of B and C is the same and the output of A and D is the same due to the incident radiation on the photodiode $B=C$ and $A=D$. Figure (9-c) shows the output of the four photodiode after the inverse amplifier, the positive output due to the ADC work on the positive signal. Fig. 9d) shows the output value of X and Y for this condition $Y=0$ and $X=+ve$ value due to spot center shifted in the positive X with the detector center. The fourth case for the laser spot shifted in X axis and a shift in Y axis the incident radiation on $A=2B=2C=2D$.

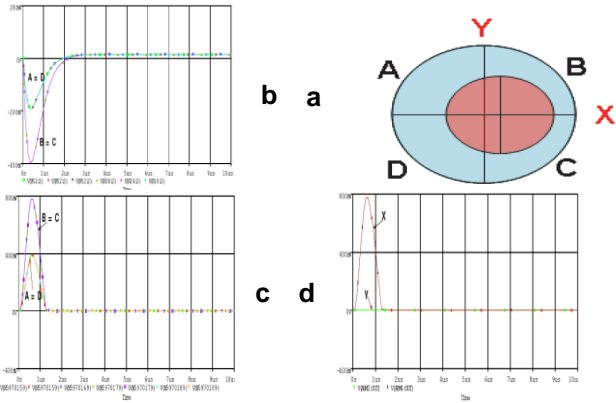


Fig. 9. laser spot shifted from the center of the quadrant detector (a), 4 PD output after amplifier (b), after inverse amplifier (c), spot position interims of X and Y (d).

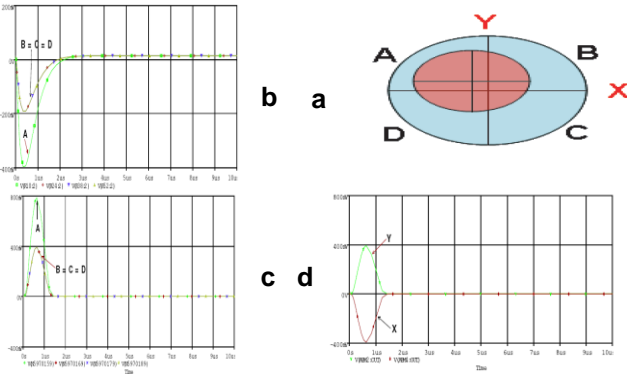


Fig. 10. Laser spot shifted from the center of the quadrant detector, 4 PD output after amplifier (b), after inverse amplifier (c), spot position interims of X and Y (d).

Fig. 10a) shows the laser spot with the quadrant detector the laser spot shifted from the center of the quadrant detector in positive X-axis and zero shift in the Y-axis. Starting with the spot shifted in X axis from the center of the quadrant detector, in this case, the illuminated areas of the four photodiodes are not the same and follow the condition $B=C=2A=2D$, then the represented value for the incident radiation is not the same for the four photodiodes. Figure (10-b) shows the output of the four photodiode after the amplifier the output of B and C is the same and the output of A and D is the same due to the incident radiation on the photodiode $B=C$ and $A=D$. Fig. 10c) shows the output of the four photodiode after the inverse amplifier, the positive output due to the ADC work on the positive signal. for the position information, we take the output from ADC and make processing on it to calculate the Y and X equations for laser

spot determination according to equation (1) and (2). Fig. 10-d) shows the output value of X and Y for this condition $X=-ve$ and $Y=+ve$ value due to spot center shifted in the negative X and +ve Y axis with the detector center.

We conclude that according to the value of incident radiation and the illuminating areas of photodiode the center of the spot can be evaluated with respect to quadrant detector center.

IV. HARDWARE IMPLEMENTATION

Fig. 11 shows the routing file and the designed PCB card and also the distribution of the components on the two cards which contain the schematic from the preamplifier to the ADC.

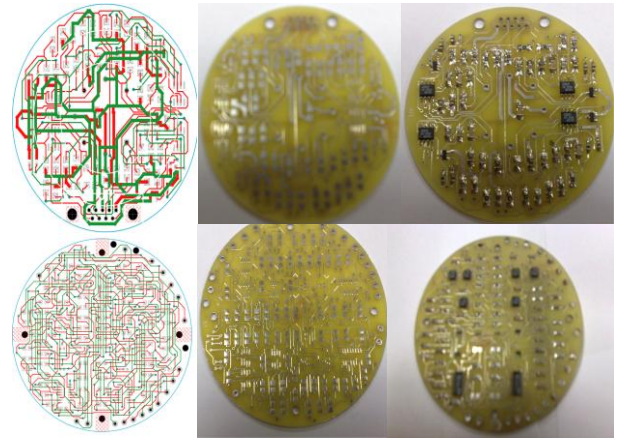


Fig. 11. Routing fill, PCB design, component distribution.

Fig. 12 shows the microcontroller card that used to perform processing on the data output from ADC. A microcontroller was used to perform processing on the signal and calculate the spot position on the detector.

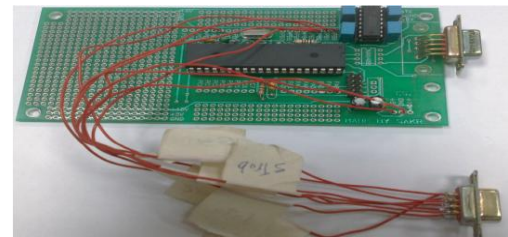


Fig.12. Microcontroller card used to make processing on the data output from ADC .

Fig. (1) shows the assembled PCB card with Quadrant detector system.



Fig. 13. Assembled system with the necessary optical system.

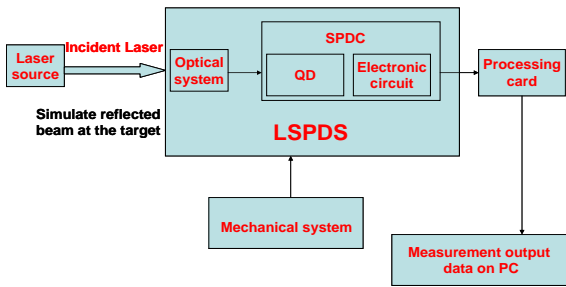


Fig. 14. General block diagram for the laser spot position determination measurement

Fig. 14 shows the general block diagram for laser spot position determination system containing laser source that simulating the incident radiation, an optical system to collimate incident radiation on the quadrant detector, microcontroller to make processing and calculation, the mechanical system to adjust seeker at different angles.

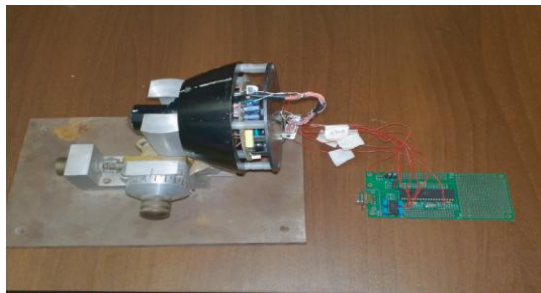


Fig. 15. Complete system including the microcontroller card .

Fig. 15 shows the complete system including the microcontroller card for the processing on the output signal to determine the spot position. A microcontroller was used to perform processing on the signal and calculate the spot position on the detector. We put the seeker on the moving mechanical system to be adjusted at different angles in the direction of the laser, with the help of the processed program we calculated the spot position on the detector.

V. SIMULATION AND RESULTS

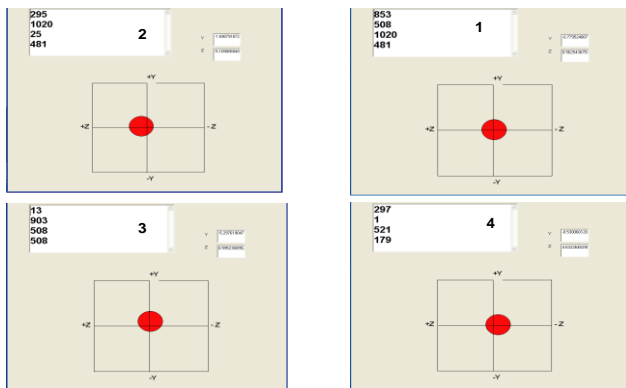


Fig. 16. Calculation of the spot position at the detector with the seeker adjusted at a different angle of Z and Y.

Fig. 16.1) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=0$ and $Z=0$. From Fig. 16.1 we conclude that when the seeker positioned at $Y=0$ degree and $Z=0$ degree, the measurement of the spot on the detector $Y=0.7$ degree and $Z=0.5$ degree has a good agreement with the seeker position and the small difference

value is due to the error in the mechanical system and the stability of the available laser source. Fig. 16.2) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=0$ and $Z=5$. From Fig. 16.2) we conclude that when the seeker positioned at $Y=0$ degree and $Z=5$ degree, the measurement of the spot on the detector $Y=0.5$ degree and $Z=4.63$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source. Fig. 16.3) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=5$ degree and $Z=0$ degree. From Fig. 16.3) we conclude that when the seeker positioned at $Y=5$ degree and $Z=0$ degree, the measurement of the spot on the detector $Y=5.29$ degree and $Z=0.59$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source. Fig. 16.4) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=0$ degree and $Z=-5$ degree. From Fig. 16.4) we conclude that when the seeker positioned at $Y=0$ degree and $Z=-5$ degree the measurement of the spot on the detector $Y=0.5$ degree and $Z=-4.63$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source.

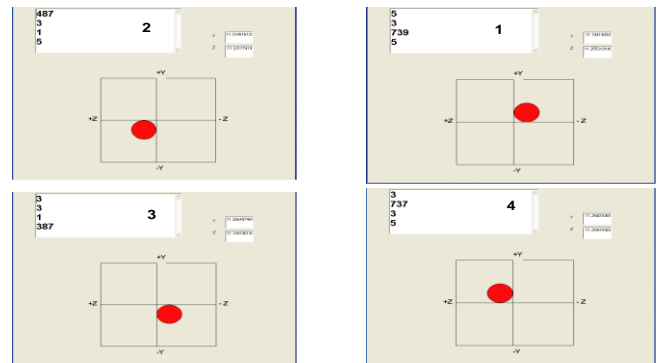


Fig. 17. Calculation of the spot position at the detector with the seeker adjusted at a different angle of Z and Y.

Fig. 17.1) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=10$ degree and $Z=-10$ degree. From Fig. 17.1) we conclude that when the seeker positioned at $Y=10$ degree and $Z=-10$ degree, the measurement of the spot on the detector $Y=11.2$ degree and $Z=-11.2$ degree has a good agreement with the seeker position and the small difference value is due to

the error in the mechanical system and the stability of the available laser source. Fig. 17.2) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=-10$ degree and $Z=10$ degree. From Fig. 17.2) we conclude that when the seeker positioned at $Y=-10$ degree and $Z=10$ degree the measurement of the spot on the detector $Y=-11.3$ degree and $Z=11.2$ degree, has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source. Fig. 17.3) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=-10$ degree and $Z=-10$ degree. From Fig. 17.3) we conclude that when the seeker positioned at $Y=-10$ degree and $Z=-10$ degree, the

measurement of the spot on the detector $Y=-11.26$ degree and $Z=-11.2$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source. Fig. 17.4) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=10$ degree and $Z=10$ degree. From Fig. 17.4) we conclude that when the seeker positioned at $Y=10$ degree and $Z=10$ degree the measurement of the spot on the detector $Y=11.2$ degree and $Z=11.2$ degree, has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source.

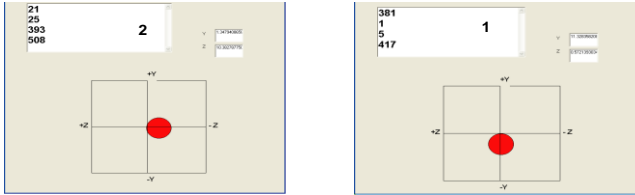


Fig. 18. The calculation of the spot position at the detector with the seeker adjusted at a different angle of Z and Y.

Fig. 18.1) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=-10$ degree and $Z=0$ degree. From Fig. 18.1) we conclude that when the seeker positioned at $Y=-10$ degree and $Z=0$ degree, the measurement of the spot on the detector $Y=-11.3$ degree and $Z=0.57$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source. Fig. 18.2) shows the calculation of the spot position at the detector with the seeker adjusted at $Y=0$ degree and $Z=-10$ degree. From Fig. 18.2) we conclude that when the seeker positioned at $Y=0$ degree and $Z=-10$ degree, the measurement of the spot on the detector $Y=1.3$ degree and $Z=-10.4$ degree has a good agreement with the seeker position and the small difference value is due to the error in the mechanical system and the stability of the available laser source.

Fig. 16 – 18 a be summarized in Table I.

TABLE I: THE ADJUSTED AND MEASUREMENT DATA OF THE DETECTOR POSITION .

angle (degree)	adjusted Y	adjusted Z	measured Y	measured Z	Y error	Z error
0	0	0	0.7	0.5	0.7	0.5
5	0	5	0.5	4.63	0.5	0.37
5	5	0	5.29	0.59	0.29	0.59
5	0	-5	0.5	-4.63	0.5	-0.37
5	5	5	4.9	6.1	0.1	1.1
10	10	10	11.2	11.2	1.2	1.2
10	-10	-10	-11.26	-11.2	-1.2	-1.2
10	10	-10	11.2	-11.2	1.2	-1.2
10	-10	10	-11.3	11.2	-1.2	1.2
10	0	-10	1.3	10.4	1.3	-0.4
10	-10	0	11.3	0.57	-1.3	0.57

We conclude that at any change of the seeker position the spot position on the detector changes and from all figures the spot coincides with the measurement ranging from 93% to 97%. The error in value is attributed to the error in the moving mechanical system used to adjust seeker to different angles and the effect of the field of view of the optical

system.

VI. CONCLUSION

LSPDC including QD, preamplifier, post amplifier and ADC was analyzed and modeled. The LSPDC was implemented in ORCAD pspice tool and then implemented in hardware. The validation of the model using a proposed setup was performed using a laser source simulating the incident radiation and a mechanical system to enable adjusting different angles; good agreement was achieved between the measured and simulated results.

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REFERENCES

- [1] M. F. Heweage, A. M. Mokhtar, and M. A. Soliman, "Modeling and performance analysis of a quad detector," ASAT-14, 2011.
- [2] G. Li, L. Li, H. B. Shen, W. S. Hua, S. J. Mao, and Y. B. Wang, "Quantitative evaluation of high repetition rate laser jamming effect on the pulsed laser rangefinder," Elsevier GmbH. All rights reserved.
- [3] R. J. Perry and K. Arora. (1996). Using pspice to simulate the photoresponse of ideal CMOS integrated circuit photodiodes. [Online]. Available: http://en.wikipedia.org/wiki/Operational_amplifier_applications.
- [4] L. H. Zhang, F. Q. Wang, B. Tang, L. J. Sun, and X. H. Ma, "Research on the tracking error of a laser tracking system," *Lasers in Eng.*, vol. 26, pp. 391–399, 2013.
- [5] Z. P. Barbaric1 and L. M. Manojlovic, "Optimization of optical receiver parameters for pulsed laser tracking systems," *Microwave Review*, September 2003.
- [6] L. G. Kazovsky, Theory of tracking accuracy of laser systems," *Optical Engineering*, vol. 22, no. 3, p. 339, 1983.
- [7] E. O. Doebelin, "Measurement systems application and design," Mc Graw-Hill, 1990.
- [8] K. A. M. Toyoda and Y. Suzuki, "Measurement of the characteristics of a quadrant avalanche photodiode and its application to a laser tracking system," *Optical Engineering*, vol. 41, no.1, January 2002.
- [9] C. Wachten, L. Friedrich, C. Müller, H. Reinecke, and C. Ament, "Sensor array for tracking systems," in *Proc. SPIE 6716*, 2007.
- [10] Y. Liu, D. Xu, and M. Tan, "A new pre-alignment approach based on four-quadrant-photo-detector for IC mask," *Int. J. Autom. Comput.*, 2007.
- [11] D. Decoster and J. Harari, *Optoelectronic Sensors*, 2009.
- [12] S. Rousseau. (2003). Laser-guided missiles an accurate and efficient Weapon. [Online]. Available: http://www.nd.edu/_techrev/Archive/Spring2002/a9.html
- [13] N. Gayathri, "Laser-based guidance system for projectile," *IOSR Journal of Engineering*, 2015.



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